# Adaptive Demodulation for Wireless Systems in the Presence of Frequency-Offset Estimation Errors

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April 18, 2018 Honolulu, HI INFOCOM 2018







Supporting higher data rates and better spectral efficiency

□ mmWave systems: up to 64-QAM



Source: [http://www.profheath.org/research/millimeter-wave-cellular-systems/]

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Providing higher security by obfuscating payload's modulation scheme Payload's modulation order leaks the payload size and data rate Used to launch various attacks: user tracking, traffic analysis, selective jamming, etc.

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Example (802.11a systems):

- 1 ms **QPSK**-modulated payload:
  - $\rightarrow$  250 OFDM symbols (symbol duration: 4µs)
    - $\rightarrow$  24 Mbps data rate (2 bits/symbol, 48 subcarriers)
      - $\rightarrow$  3,000 coded bytes



## Review: Modulation Obfuscation [MobiHoc'15, TIFS'16]

## Hide (obfuscate) the payload's modulation scheme

Covertly "embed" modulated symbols of every payload into the dense constellation map of the highest-order modulation scheme

Hide true modulation scheme without changing it

→ Same information rate (rate adaptation algorithm works as normal)

> Example:



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# Challenge: High Sensitivity to Phase Offset

Denser constellation maps are more vulnerable to phase offset Example: BPSK vs. 16-QAM at the receiver



## **Residual Phase Offset**

Common causes

Imperfect channel estimation

#### **Residual carrier frequency offset (CFO)**

CFO: Mismatch between operating freq. of two devices (+ Doppler shift)

## Receiver uses frame preamble to estimate CFO



In a noisy channel, CFO estimation can never be perfect

 $\Delta \varphi(t) = 2\pi \times \delta_f \times t$ time Time-varying phase offset **Residual CFO** 

# Symbol Distribution under CFO-induced Phase Offset

Symbols with unequal amplitudes have unequal sensitivity to  $\Delta \varphi(t)$ Example: Heatmap of the location of two received 16-QAM symbols under AWGN



## Theoretical Analysis (Wi-Fi Systems)

Probability density function of phase offset per symbol (under AWGN)



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## **CFO-Aware Demodulation Boundaries**

Default boundaries vs. CFO-aware boundaries



 $(4 \times 2)$  8-APSK

BER Performance gain Example: 16-QAM



## (Uncoded) Modulation Obfuscation

Map symbols of a mod. scheme to a subset of higher-order symbols Selection of an optimal sub-constellation is based on a secret *j* Example: QPSK → 16-QAM



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# **Modulation Obfuscation under Phase Offset**

Denser constellation maps increase the vulnerability to phase offset

 $\rightarrow$  A low-order modulation scheme becomes more sensitive to CFO than usual

Example:



No residual CFO



With residual CFO





## **CFO-Aware Demodulation for Uncoded Obfuscation**

Default demodulataion boundaries may not be optimal Example: mapping BPSK/QPSK symbols to (a subset of) 16-QAM symbols





## Coded Modulation Obfuscation [TIFS'16]

Use lightweight Trellis-Coded Modulation (TCM) to improve BER Needs (at least) two optimal sub-constellations

### Example:

QPSK  $\rightarrow$  16-QAM using TCM

Needs two sub-constellations ( $2 \times 4$  16-QAM symbols) for the four QPSK symbols



Similarly, two sub-constellation ( $2 \times 2$  16-QAM symbols) for the two BPSK symbols

Covertly vary the sub-constellations based on secret *j* to cover all possible symbols



## Optimizing Coded Obfuscation w/ Phase Offset Consideration

1) Find optimal pairs of sub-constellations with maximum inter-subconstellation distance



- 1) What if there are multiple optimal pairs?
- $\Phi$ : max phase offset 00 00000O O0 0 0 0- O - 🔴 000  $\odot$ O = O $\bullet \circ \circ \circ \circ \circ$ 0 0 0 0 0 0 0 0 0 O O $\bigcirc$  $\odot$  $\bigcirc$   $\bigcirc$ Tie breaker:  $\Phi$ 0000000 0 0 0 0 0 0 0 0 0 0000 0 O = O0000 00000000 00 Vertex cover algorithm  $\odot$ 0 0 0 0 00  $\odot$ 0 0 0 000 0000 0000 00 Ex: BPSK  $\rightarrow$  64-QAM 0000000 00000 00000000 000000 0000000  $\circ \circ \circ \bullet \circ \circ \circ \circ$ 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0  $\Phi = 0.82$  $\Phi = 0.66$  $\Phi = 0.64$

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# Performance (Gain) Improvement

Least-complex TCM is sufficient to maintain the performance under AWGN 1) Mapping to 16-QAM

	Minimum-distance reduction after mapping to 16-QM	Gain (uncoded)	Gain (w/ 2-state TCM)	Enhanced Gain (w/ 2-state TCM)
BPSK	2 → 1.79	−0.97 dB	−0.46 dB	0.79 dB
QPSK	$1.41 \rightarrow 1.26$	-0.97dB	0 dB	0.79 dB

#### 2) Mapping to 64-QAM

	Minimum-distance reduction after mapping to 64-QM	Gain (uncoded)	Gain (w/ 2-state TCM)	Enhanced Gain (w/ 2-state TCM)
BPSK	$2 \rightarrow 1.75$	$-1.18~\mathrm{dB}$	—1.05 dB	0 dB
QPSK	$1.41 \rightarrow 1.23$	$-1.18~\mathrm{dB}$	—0.92 dB	0.58 dB
16-QAM	$0.63 \rightarrow 0.62$	-0.21 dB	0.76 dB	1.55 dB

## **Resulting Robustness to Phase Offset**

What is the maximum phase offset (in Rad) that does not create error?

	Default	→ 16-0	AM	→ 64-QAM	
		CBM scheme	Proposed	CBM scheme	Proposed
BPSK	$\pi/2 = 1.57$	0.295	0.545	0.135	0.51
QPSK	$\pi/4 = 0.78$	0.295	0.464	0.135	0.381
16-QAM	0.259	N/A	N/A	0.135	0.165

BER performance under phase offset
Example:
QPSK → 16-QAM

 $QPSK \rightarrow 64-QAM$ 



## Conclusions

Sensitivity of higher-order modulation schemes to phase offset may hinder using (and securing) them in emerging wireless systems

Default demodulation boundaries are inept at high transmission rates (i.e., at dense constellation maps)

Adaptive (CFO-aware) demodulation boundaries can achieve up to 2 dB gain for 16-QAM and 64-QAM modulation schemes

By redesigning the coding scheme for modulation obfuscation w.r.t. phase offset, one can achieve additional 2 - 3 dB gain

 $\rightarrow$  Up to 5 *dB* gain for modulation obfuscation over conventional demodulation schemes that are not obfuscated and are oblivious to residual CFO